



AL58263

16-CHANNEL, 16-BIT GRAYSCALE PWM DIMMING LED DRIVER

Description

The AL58263 is a 16-channel constant-current LED driver with 16-bit grayscale Adaptive Pulse Density Modulation (APDM) and supports error diagnostics, power-saving functionality, and current gain control. This distinctive APDM technology abates the non-ideal IOUT distortion due to non-symmetric transient responses and enhances the refresh rate by efficiently separating the frame waveform.

The device operates over a 3V to 5.5V input voltage range, 15V output channel voltage, fast 25MHz DCK input, and delivers up to 55mA of high-accuracy current to each LED string. Each channel's output current can be programmable through a digital interface individually. All channels' output current can be set by one external sensing resistor, along with a 6-bit global current control register.

The AL58263 has built-in diagnostics and LED open/short protection, including error detection. These error results are stored in a register for the MCU to read out.

The AL58263 is available in the TSSOP-24EP (Type A1-B) package and specified over the -40°C to +85°C ambient temperature range.

Features

- Input Voltage VDD: 3V to 5.5V
- Output Current Range
 - 2~55mA/5V, 2~35mA/3.3V
 - ±0.1% Output current regulation capability
 - 6-bit global current control: from 12.5% to 200%
- 16 Constant-Current Sink Output Channel
 - 16-bit grayscale resolution with Adaptive Pulse Density Modulation control
 - 15V rated output channels for long LED strings
 - Fast Transient Response supports external grayscale clock with double edge up to 16MHz
 - ±1.5% (typ.) LED Current accuracy between channels
 - ±3% (typ.) LED Current accuracy between chips
 - Non-scramble waveform for high power LED application
 - Grayscale counter reset selection
 - Grayscale data synchronization selection
- · Diagnosis and Protections
 - Error detection includes LED Open, LED Short, Output port leakage, Output short-to-GND, Output short-to-Power and REXT short-to-GND
 - Error detection LEDs on at 0.1mA to avoid any flickering
 - Short detection threshold voltage selection (2/3/4/4.5V)
 - Sleep and 0-data mode to lower down the supply current
 - Pre-overtemperature warning
- 4-Wire Serial Interface (LAT, DI, DO, DCK)
 - 25MHz clock frequency for data transfer
 - · EMI reduction grayscale clock
 - Cascaded capability (Max 1,440 devices)
 - External GCK watchdog
 - Stagger outputs delay for EMI reduction
- Totally Lead-Free & Fully RoHS Compliant (Notes 1 & 2)
- Halogen and Antimony Free. "Green" Device (Note 3)
- For automotive applications requiring specific change control (i.e. parts qualified to AEC-Q100/101/104/200, PPAP capable, and manufactured in IATF 16949 certified facilities), please contact us or your local Diodes representative. https://www.diodes.com/quality/product-definitions/

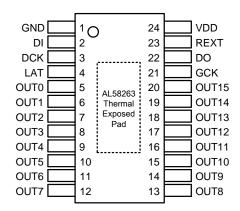
Pin Assignments

(3D Photo - Not to Scale)



(3D Step file available upon request)

(Top View - Not to Scale)



TSSOP-24EP (Type A1-B)

Applications

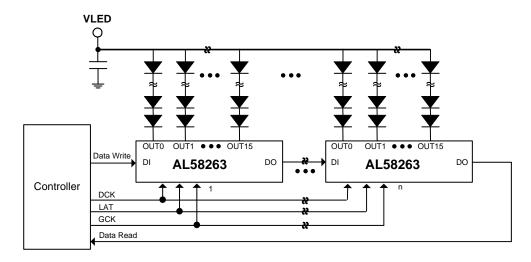
- Indoor and outdoor LED video displays
- Variable message signs (VMS)
- Traffic signs
- Outdoor billboard signage
- LCD display backlighting

Notes:

- 1. No purposely added lead. Fully EU Directive 2002/95/EC (RoHS), 2011/65/EU (RoHS 2) & 2015/863/EU (RoHS 3) compliant.
- 2. See https://www.diodes.com/quality/lead-free/ for more information about Diodes Incorporated's definitions of Halogen- and Antimony-free, "Green" and Lead-free.
- 3. Halogen- and Antimony-free "Green" products are defined as those which contain <900ppm bromine, <900ppm chlorine (<1500ppm total Br + Cl) and <1000ppm antimony compounds.



Typical Applications Circuit

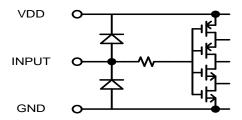


Pin Descriptions

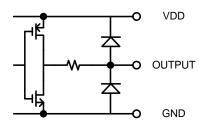
	PIN	
NAME	TSSOP-24EP (Type A1-B)	DESCRIPTION
GND	1	Ground terminal
DI	2	Serial data input terminal.
DCK	3	Synchronous clock input terminal for serial data transfer. Data is sampled at the rising edge of DCK.
LAT	4	Input terminal of data strobe and mode setting. Combine DCK with LAT signal to execute the mode control
OUT0 to OUT15	5 to 20	Sink constant-current outputs (open-drain).
GCK	21	External double-edge grayscale clock input for APDM operations
DO	22	Serial data output terminal.
REXT	23	External resistors connected between REXT and GND for output current value setting.
VDD	24	Supply voltage terminal.

Equivalent Circuit of Inputs and Outputs

1. DCK, DI, LAT, GCK terminals

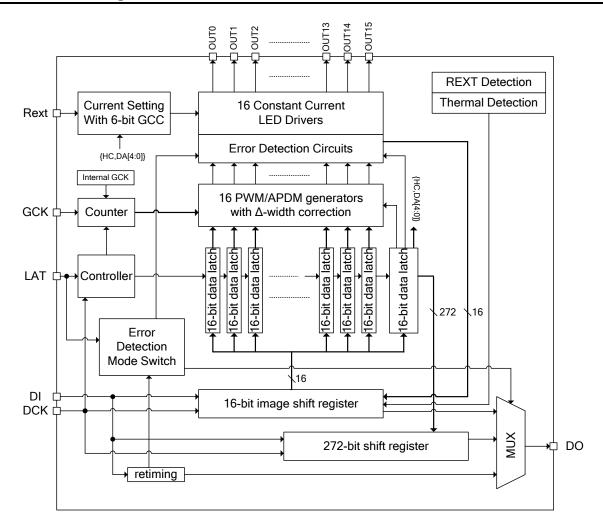


2. DO terminal





Functional Block Diagram





Absolute Maximum Ratings (@TA = +25°C, unless otherwise specified.)

Symbol	Parameters	Ratings	Unit
VDD	Supply Voltage	-0.3 ~ 7.0	V
$V_{DI}, V_{DCK}, V_{GCK}, V_{LAT}$	Input Voltage	-0.3 ~ VDD+0.3	V
IOUT	Output Current	60	mA
VOUT	Output Voltage	-0.3 ~ 16	V
FDCK	Input Clock Frequency	25	MHz
FGCK	Input Grayscale Frequency	16	MHz
IGND	GND Terminal Current	960	mA

⁽¹⁾ Stresses above these ratings may cause permanent damage. Exposure to absolute maximum conditions for extended periods may degrade device reliability. These are stress ratings only and functional operation of the device at these or any other condition beyond those specified is not supported.

ESD Ratings

Symbol	Parameter	Value	Unit
1/	Human-Body Model (HBM)	2000	V
VESD	Charged-Device Model (CDM)	750	V

Package Thermal Data (Note 4)

Package	θ _{JC} Thermal Resistance	θ _{JA} Thermal Resistance	P _{DIS}
	Junction-to-Case (°C/W)	Junction-to-Ambient (°C/W)	T _A = +25°C, T _J = +105°C
TSSOP-24EP (Type A1-B)	15	28	2.86

Note:

Recommended Operating Conditions (@TA = +25°C, unless otherwise specified.)

Symbol	Parameter	Min	Тур	Max	Unit
V_{DD}	Operating Supply Voltage	3.0	3.3	5.5	V
l _{оит}	Output Current (Note 5)	2	_	55	mA
V _{оит}	Output Voltage	1	_	15	V
T _A	Ambient Temperature (Note 5)	-40	_	+85	°C
TJ	Junction Temperature	-40	_	+125	°C
Tstg	Storage Temperature	-55	_	+150	°C

Note:

5. Dependent on ambient temperature, LED voltage, package thermal limitation, and PCB layout.

⁽²⁾ All voltage values are with respect to ground terminal.

⁽³⁾ Operation at Tj(max) = 150°C may reduce the device reliability; therefore, the suggested junction temperature of the device is under 125°C.

^{4.} Test condition: Device mounted on FR-4 PCB (51mm x 51mm 2oz copper, minimum recommended pad layout on top layer and thermal vias to bottom layer with maximum area ground plane. For better thermal performance, larger copper pad for heat-sink is needed.



Electrical Characteristics (VDD=5.0V @TA = +25°C, unless otherwise specified.)

CHARACTERISTIC	SYMBOL	CONDITION	MIN.	TYP.	MAX.	UNIT
Input Voltage "H" Level	VIH	CMOS logic level	0.7VDD	_	VDD	V
Input Voltage "L" Level	VIL	CMOS logic level	GND	_	0.3VDD	V
Output Leakage Current	ILK	VOUT = 15 V	_	_	0.1	uA
Delicina Comment (DO)	IOL	VOL = 0.4V	1	3.4	5	A
Driving Current (DO)	IOH	VOH= VDD-0.4	-5	-3.4	-1	mA
Output Current Skew (Channel-to-Channel) (Note 6)	dIOUT1	VOUT = 1.0 V Rrext = 560 Ω	_	±1.5	±3	%
Output Current Skew (Chip-to-Chip) (Note 7)	dIOUT2	GCC=101011	_	±3	±6	%
Output Current Skew (Channel-to-Channel) (Note 6)	dIOUT3	VOUT = 1.0 V	_	±1.5	±3	%
Output Current Skew (Chip-to-Chip) (Note 7)	dIOUT4	Rrext = 7 KΩ GCC=101011	_	±3	±6	%
Output Voltage Regulation (Note 8)	% / VOUT	Rrext = 560Ω VOUT = $1 \text{ V} \sim 3 \text{ V}$	_	±.0.1	±.1	% / V
Supply Voltage Regulation (Note 9)	% / VDD	Rrext = 560Ω VDD = $3 \text{ V} \sim 5.5 \text{ V}$	_	±0.6	±1	% / V
	I _{DD1(off)}	Rrext = $7 \text{ K}\Omega$ all outputs turn off	1	2.5	4	
Supply Current (Note 10)	I _{DD2(on)}	Rrext = $7 \text{ K}\Omega$ all outputs turn on	1	2.5	4	mA
Supply Current (Note 10)	I _{DD3(off)}	Rrext = 560Ω all outputs turn off	1	5.5	7.5	MA
	I _{DD4(on)}	Rrext = 560Ω all outputs turn on	1	5.5	7.5	
Thermal alarm temperature	TAL	_	_	+155		°C
Thermal alarm temperature hysteresis	TAL_HYS	_	_	+15	_	°C

Notes: 6. Channel-to-channel skew is defined by the formula below:

$$\Delta(\%) = \left[\frac{Iout_n}{(Iout_0 + Iout_1 + ... + Iout_{15})} - 1 \right] * 100\%$$

7. Chip-to-Chip skew is defined by the formula below:

$$\Delta(\%) = \left[\begin{array}{c|c} (\underbrace{Iout_0 + Iout_1 + \ldots + Iout_{15}}) - (Ideal \ Output \ Current) \\ \hline (Ideal \ Output \ Current) \end{array}\right] * 100\%$$

8. Output voltage regulation is defined by the formula below:

$$\Delta \left(\% \ / V \right) = \left[\begin{array}{c} Iout_n(@Vout_n = 3V) - Iout_n(@Vout_n = 1V) \\ Iout_n(@Vout_n = 3V) \end{array} \right] * \frac{100\%}{3V - 1V}$$

9. Supply voltage regulation is defined by the formula below:

$$\Delta \left(\% \, / V \right) = \left[\begin{array}{c} Iout_n(@V_{DD} = 5.5V) - Iout_n(@V_{DD} = 3V) \\ Iout_n(@Vcc = 3V) \end{array} \right] * \frac{100\%}{5.5V - 3V}$$

10. IO excluded.



Electrical Characteristics (VDD=3.3V @T_A = +25°C, unless otherwise specified.)

CHARACTERISTIC	SYMBOL	CONDITION	MIN.	TYP.	MAX.	UNIT
Input Voltage "H" Level	VIH	CMOS logic level	0.7VDD	_	VDD	V
Input Voltage "L" Level	VIL	CMOS logic level	GND	_	0.3VDD	V
Output Leakage Current	ILK	VOUT = 15 V	_	_	0.1	uA
Division Comment (DC)	IOL	VOL = 0.4V	1	2.5	5	0
Driving Current (DO)	IOH	VOH= VDD-0.4	-5	-2.5	-1	mA
Output Current Skew (Channel-to-Channel) (Note 11)	dIOUT1	VOUT = 1.0 V Rrext = 560 Ω	_	±1.5	±3	%
Output Current Skew (Chip-to-Chip) (Note 12)	dIOUT2	GCC=101011	_	±3	±6	%
Output Current Skew (Channel-to-Channel) (Note 11)	dIOUT3	VOUT = 1.0 V	_	±1.5	±3	%
Output Current Skew (Channel-to-Channel) (Note 12)	dIOUT4	Rrext = 7 KΩ GCC=101011	_	±3	±6	%
Output Voltage Regulation (Note 13)	% / VOUT	Rrext = 560 Ω VOUT = 1 V ~ 3 V	_	±.0.1	±1	07.777
Supply Voltage Regulation (Note 14)	% / VDD	Rrext = 560 Ω VDD = 3 V ~ 5.5 V	_	±0.7	±1	% / V
	I _{DD1(off)}	Rrext = 7 K Ω all outputs turn off	1	2.3	4	
Countly Country (Nata 45)	I _{DD2(on)}	Rrext = $7 \text{ K}\Omega$ all outputs turn on	1	2.3	4	Α
Supply Current (Note 15)	I _{DD3(off)}	Rrext = 560Ω all outputs turn off	1	4.7	7.5	mA
	I _{DD4(on)}	Rrext = 560Ω all outputs turn on	1	4.7	7.5	
Thermal alarm temperature	TAL	_	_	+155	_	°C
Thermal alarm temperature hysteresis	TAL_HYS	_	_	+15	_	°C

Notes:

11. Channel-to-channel skew is defined by the formula below:

$$\Delta(\%) = \left[\frac{Iout_n}{(Iout_0 + Iout_1 + ... + Iout_{15})} - 1 \right] * 100\%$$

12. Chip-to-Chip skew is defined by the formula below:

$$\Delta \left(\%\right) = \left[\begin{array}{c|c} \underbrace{(Iout_0 + Iout_1 + ... + Iout_{15})}_{\left(Ideal \ Output \ Current\right)} - (Ideal \ Output \ Current)}_{\left(Ideal \ Output \ Current\right)} \right] * 100\%$$

13. Output voltage regulation is defined by the formula below:

$$\Delta \left(\% \ / V \right) = \left[\begin{array}{c} Iout_n(@Vout_n = 3V) - Iout_n(@Vout_n = 1V) \\ Iout_n(@Vout_n = 3V) \end{array} \right] * \frac{100\%}{3V - 1V}$$

14. Supply voltage regulation is defined by the formula below:

$$\Delta \left(\% / V \right) = \left[\begin{array}{c} Iout_n(@V_{DD} = 5.5V) - Iout_n(@V_{DD} = 3V) \\ Iout_n(@Vcc = 3V) \end{array} \right] * \frac{100\%}{5.5V - 3V}$$

15. IO excluded.



Switching Characteristics (VDD = 5.0V @T_A = +25°C, unless otherwise specified.) (Guarantee by design)

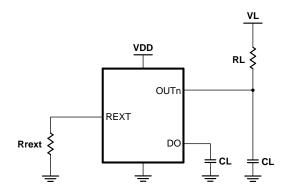
CHARA	ACTERISTIC	SYMBOL	CONDITION	MIN.	TYP.	MAX.	UNIT
Propagation Delay ('L to 'H')	DCK-DO	tpLH3		_	30	35	
Propagation Delay ('H' to 'L')	DCK-DO	tpHL3		_	30	35	
	LAT	tw _(LAT)		50	_	_	
Pulse Duration	GCK	tw _(GCK)		20	_	_	
	DCK	tw _(DCK)		20	_	_	
Satura Timo	LAT	tsu _(LAT)	\///\ \/\DD	20	_	_	
Setup Time	DI	tsu _(D)	VIH = VDD VIL = GND	5	_	_	ns
Hold Time	LAT	th _(LAT)	Rrext = 560Ω	20	_	_	
Hold Tillle	DI	th _(D)	VL =5.0 V RL = 150 Ω	5	_	_	
Hold Time of Instruction		th _(CM)	CL = 13 pF	20	_	_	
LED Error Detection Time	Э	Terr1		100	_	_	
Short-to-ground Detectio	n Time	Terr3		160	_	_	
Interval between two command pulses		tpls		230	250	_	
Wake-up time from sleep	Twake		_	_	4000	us	
Data Clock Frequency	F _{DCK}		_	_	25	NAL I-	
Grayscale Clock Frequer	ncy (double-edge)	F _{GCK}		_	_	16	MHz



Switching Characteristics (VDD = 3.3V @TA = +25°C, unless otherwise specified.) (Guarantee by design)

CHAR	ACTERISTIC	SYMBOL	CONDITION	MIN.	TYP.	MAX.	UNIT
Propagation Delay ('L to 'H')	DCK-DO	tpLH3		_	45	55	
Propagation Delay ('H' to 'L')	DCK-DO	tpHL3		_	45	55	
	LAT	tw _(LAT)		50		_	
Pulse Duration	GCK	tw _(GCK)		20		_	
	DCK	tw _(DCK)		20	_	_	
Cotus Time	LAT	tsu _(LAT)		20	_	_	
Setup Time	DI	tsu _(D)	VIH = VDD VIL = GND	5	_	_	ns
Hold Time	LAT	th _(LAT)	Rrext = 560Ω	20	_	_	
noid Time	DI	th _(D)	VL = 5.0 V $RL = 150 \Omega$	5	_	_	
Hold Time of Instruction		th _(CM)	CL = 13 pF	20	_	_	
LED Error Detection Time	e	Terr1		100	_	_	
Short-to-ground Detection	Short-to-ground Detection Time			180	_	_	
Interval between two command pulses		tpls		230	250	_	
Wake-up time from sleep or 0-data mode		Twake		_	_	4000	us
Data Clock Frequency	F _{DCK}		_	_	15	NAL I-	
Grayscale Clock Frequen	ncy (double-edge)	F _{GCK}		_		16	MHz

Switching Characteristics Test Circuit



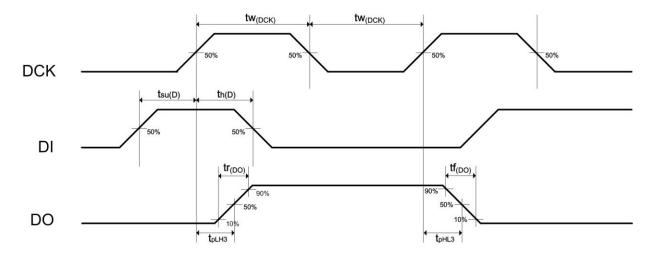
Switching Characteristics Test Circuit



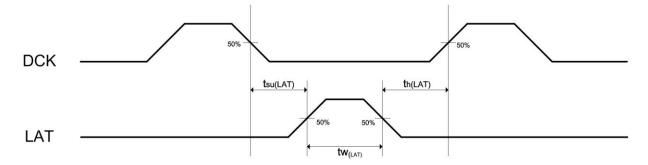
Timing Characteristics (@T_A = +25°C, unless otherwise specified.)

Timing Diagram

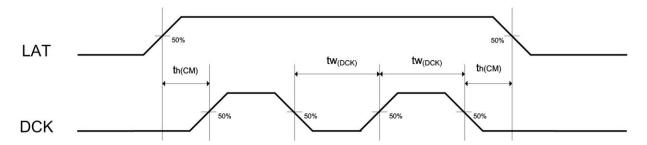
1. DCKI-DI, DO



2. GCK-LAT

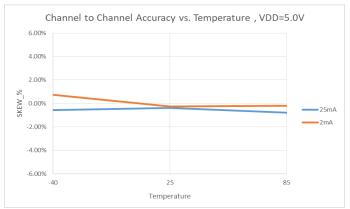


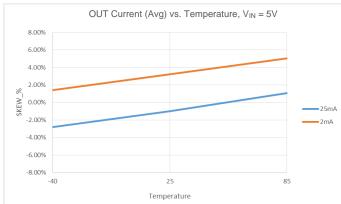
3. LAT-DCK (Instruction)





$\textbf{Typical Performance Characteristics} \ (V_{DD} = 5V, \ -40 ^{\circ}\text{C} < T_{A} < +85 ^{\circ}\text{C}, \ unless \ otherwise \ specified.})$



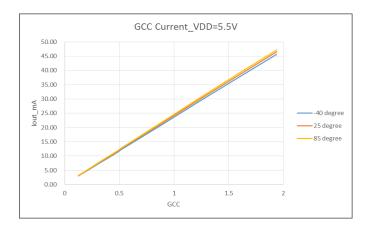


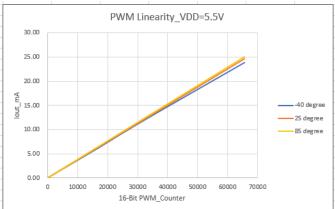
 $V_{DD} = 5V$, $I_{OUT} = 25mA$, 2mA

Figure 1. Channel to Channel Accuracy vs. Temperature

 $V_{DD} = 5V$, $I_{OUT} = 25mA$, 2mA

Figure 2. Chip to Chip Accuracy vs. Temperature





 $V_{DD} = 5V$, $I_{OUT} = 25mA$

Figure 3. Gain Control Current vs. Temperature

 $V_{DD} = 5V$, $I_{OUT} = 25mA$

Figure 4. PWM Linearity vs. Temperature



Functional Descriptions

Fast Transient Response

The AL58263 supports the fast transient response to make high image resolution possible. The GCK period of 50ns is enough to get a complete Vout waveform.

Stagger Outputs Delay

Large in-rush currents will be induced when the system activates all the outputs at once. To reduce this interference of EMI, the AL58263 is designed to have a constant length of delay time (around 16ns) between two output groups. The first group is OUT2n and the second group is OUT2n+1 (n=0~7).

Global Current Control (GCC)

The AL58263 provides 6bits global current control (GCC) which could adjust the output current by 64 steps.

The GCC bits are included in the command data defined as follow:

F	Е	D	С	В	Α	9	8	7	6	5	4	3	2	1	0
_	_					НС	DA4	DA3	DA2	DA1	DA0				_

The relationship between the Gain value and GCC bits is described below:

HC=1, Gain $= (3 \times D + 33) / 65 \times (1-0.001 \times D)$, the gain range is from 0.508 to 1.878

HC=0, Gain $= (3 \times D + 32) / 256 \times (1-0.0006 \times D)$, the gain range is from 0.125 to 0.479

Where D = DA4 x 2^4 + DA3 x 2^3 + DA2 x 2^2 + DA1 x 2^1 + DA0 x 2^0

For example, HC=1 and DA[4:0]=11000

 $D = 1x 2^4 + 1 x 2^3 + 0 x 2^2 + 0 x 2^1 + 0 x 2^0 = 24$

Gain = (3 x 24 + 33) / 65 x (1-0.001 x 24) = 1.577

Reference Resistor

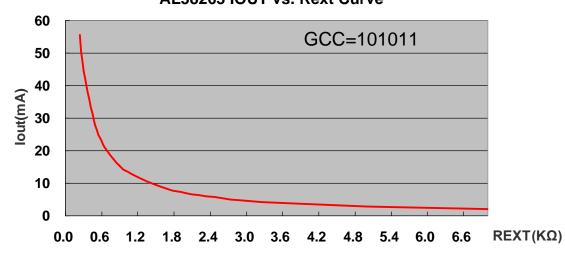
The constant current values are determined by an external resistor placed between REXT pin and GND pin. The following formula is utilized to calculate the current value:

lout (mA)
$$=$$
 13.45 / Rext(K Ω) x Gain

This 13.45 constant center at 10mA may need to be adjusted slightly base on users application and test conditions. Where Rext is a resistor placed between REXT and GND

For example, lout is around 24.12mA when Rrext= 560Ω and with GCC=101011

AL58263 IOUT vs. Rext Curve



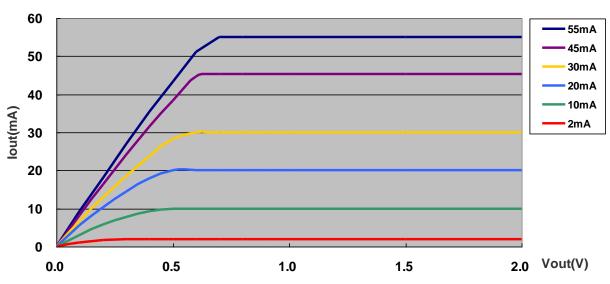
AL58263 Document number: DS46310 Rev. 2 - 2 11 of 28 www.diodes.com



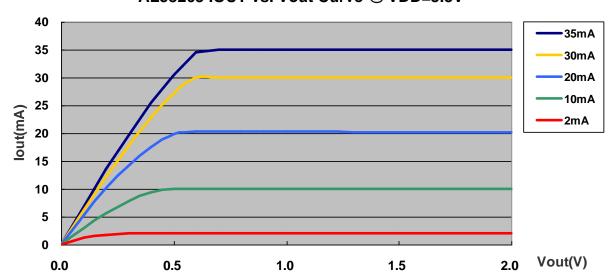
Constant-Current Output

The current characteristics could maintain invariable in the influence of loading voltage. Therefore, the AL58263 could minimize the interference of different LED forward voltages and produce the constant current. The following figures illustrate the suitable output voltage should be determined in order to keep an excellent performance.

AL58263 IOUT vs. Vout Curve @ VDD=5V

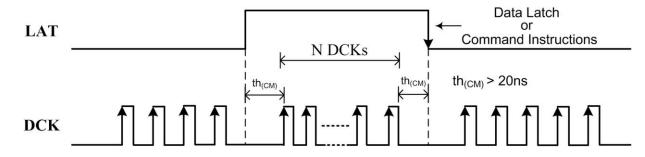


AL58263 IOUT vs. Vout Curve @ VDD=3.3V



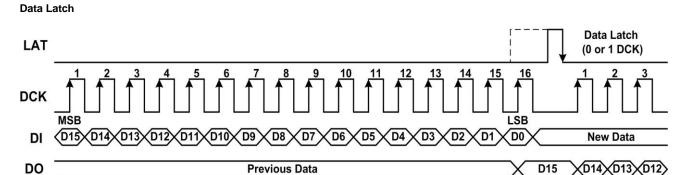


Control Instructions



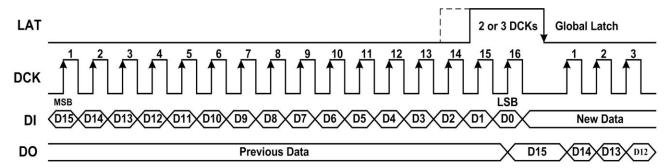
Instruction Name	Number of DCK Pulses when LAT is at High	Function Description
Data Latch	0 or 1	Strobes image data into temporary registers. In this time, constant current outputs are not updated to a new frame data.
Global Latch	2 or 3	Strobes image data from temporary registers into an APDM/PWM generator. In this time, constant current outputs are updated to a new frame data.
Reading Image Data and Command Data	6 or 7	Read out image data and command data from the latch cell to the 272bits shift registers. And the read data could shift out from DO pin by serial DCK pulses.
Return to Normal Mode	8 or 9	Switch Back from the reading data mode to the normal operating mode.
Writing Command	10 or 11	Strobes command data from the 16bits shift register into the command latch cell.
Reset APDM/PWM Counter	12 or 13	This instruction will reset APDM/PWM counter to synchronize a new frame if CMD[11]="H".



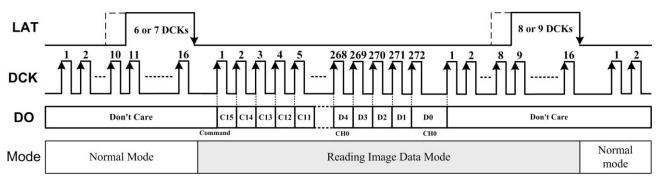


Previous Data

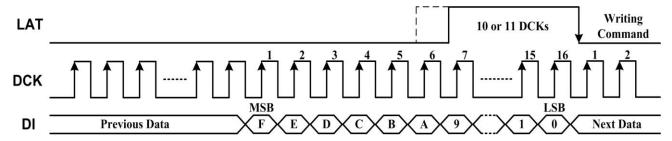
Global Latch



Reading Image Data and Command Data

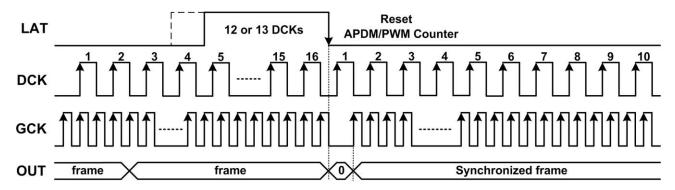


Writing Command

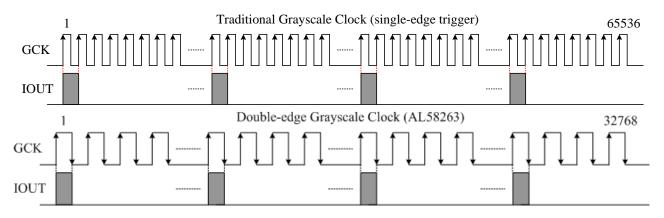




Reset APDM/PWM Counter (set CMD[11]="1")



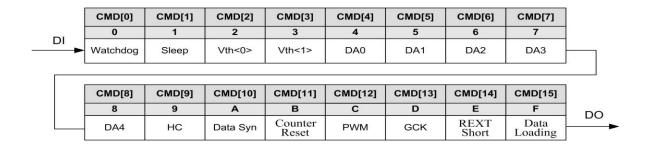
EMI Reduction Grayscale Clock (Double-edge Grayscale Clock)



A whole period of 16bits resolution must be composed by 65536 traditional grayscale clocks because constant current outputs only are triggered at the rising edge of clocks. Therefore, a controller has to transmit fast grayscale clocks in order to accomplish high refresh rate when users adopt traditional PWM chips. AL58263 supports a specific mode of double-edge grayscale clocks which trigger both at rising and falling edges of clocks. By this approach, a whole period of 16bits resolution is composed by only 32768 double-edge grayscale clocks and the electromagnetic interference would be decreased substantially due to slow grayscale clocks.



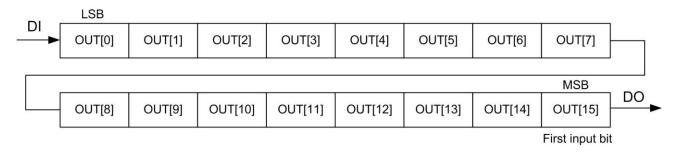
Command Data Format



CMD Bit	Initial Value	Value	Function	Description
		1'b0	Mode0	Data transmitting Format:
CMD[15]	1'b0		Modeo	15 times of data latch + one global latch
[]		1'b1	Mode1	Data transmitting Format:
				16 times of data latch + one global latch
		1'b0	Enable	When the voltage of REXT is below 0.15V,
CMD[14]	1'b0			all outputs will be forced off.
		1'b1	Disable	Disable
		1'b0	External GCK	The GCK signal stems from an external source
CMD[13]	1'b0			The GCK signal stems from an internal
52[1.0]		1'b1	Internal GCK	oscillator, 1.5MHz.
				(When CMD[13]="1",please set CMD[0]="0")
		1'b0	APDM mode	Enter the PWM mode to abate the transient
CMD[12]	1'b0		PWM mode	loss when AL58263 is as a PWM generator
		1'b1		when CMD[12]=1'b0 => APDM mode
				when CMD[12]=1'b1 => PWM mode
01.151.11	411.0	1'b0	Disable	Disable
CMD[11]	1'b0	1'b1	Enable	Execute counter reset by inserting 12 or 13
				DCKs (rising edge) when LAT keeps at high
		1'b0	Auto	When the device receives a latch signal, the new frame
CMD[10]	1'b0		Synchronization	is updated until the end of old frame
		1'b1	Manual	When the device receives a latch signal,
		6'b000000~	Synchronization	the new frame is updated immediately 6bits DA data for global current control
CMD[9:4]	6'b101011	6'b111111	G.C.C	(allow 64-step programmable current gain)
		2'b00	2.0V	(allow 04 step programmable current gain)
		2'b01	3.0V	Threshold voltage for LED short detection
CMD[3:2]	2'b00	2'b10	4.0V	(both at VDD = 5V and 3.3V)
		2'b11	4.5V	
01/5	411	1'b0	Disable	Enter 0-data power saving mode
CMD[1]	1'b0 1'b1		Enable	when 16bits PWM data of 16ch are all zeros
		1'b0	Disable	GCK watchdog
				All IOUTs will be turned off automatically when GCK
CMD[0]	1'b0	1'b1	Enable	keeps high or low level over 30ms.
		101	Enable	(Result can be read out by "Thermal Detection Mode" or
				"Rext Detection Mode")



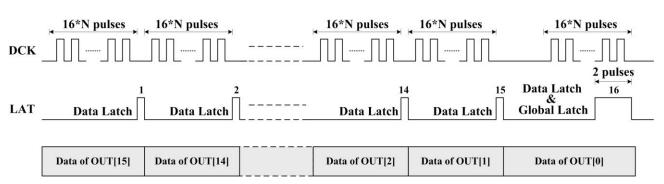
Image Data Format



16bits APDM/PWM data are transmitted into the 16bits shift register according to the format illustrated above. The first input bit is the most significant bit of each channel.

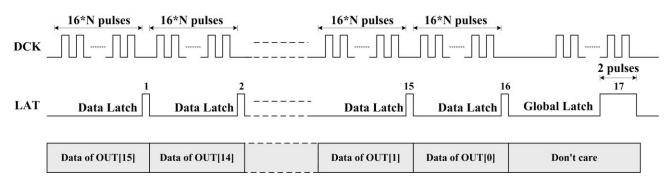
Data Transmitting Protocol

MODE0 (CMD[15]="0")



This data transmitting mode is comprised of 15 data latches and one global latch. The first data latch strobes image data for OUT[15]. And the last global latch strobes image data for OUT[0] and loads all image data of 16 channels into the APDM/PWM generator. N is the number of devices in series.

MODE1 (CMD[15]="1")

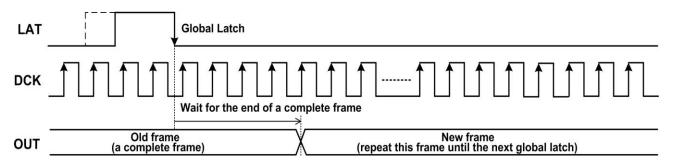


This data transmitting mode is comprised of 16 data latches and one global latch. The first data latch strobes image data for OUT[15]. And the last global latch loads all image data of 16 channels into the APDM/PWM generator but doesn't strobe any image data for output channels. N is the number of devices in series.



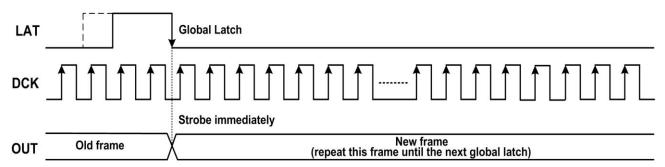
Data Synchronization

Auto Synchronization (CMD[10]="0")



AL58263 could abate the grayscale loss during the renewal of frame by this approach of auto synchronization. When the command bit CMD[10] is set to "0", this function could be executed automatically. After a global latch, all OUTs would accomplish a complete frame of old data and then update new frame data. Therefore, each frame maintains the fixed time and the assigned grayscale. A controller just provides a free-running GCK for grayscale display in this mode.

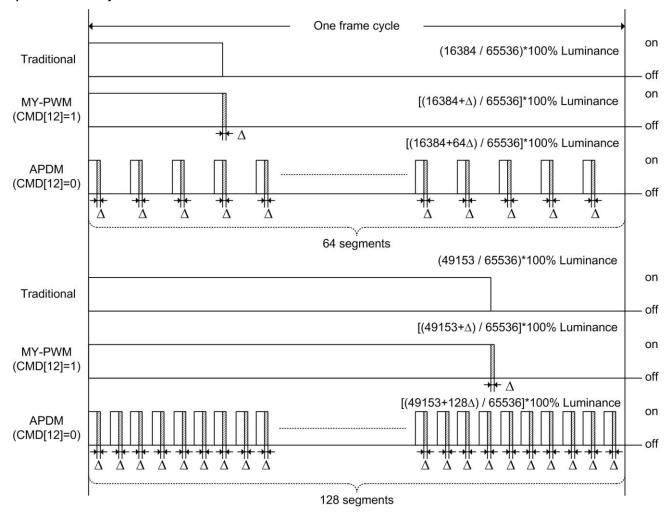




When the command bit CMD[10] is set to "1", the new frame data would be updated immediately. The old frame loses probably a little grayscale due to a non-complete frame time. In this mode, a controller must calculate carefully the number of GCKs in order to synchronize a frame.



Adaptive Pulse Density Modulation with Δ -Width Correction



Adaptive Pulse Density Modulation (APDM) is a technique to improve output current waveform distortion and increase visual refresh rate. The adaptive output waveform is determined automatically by the grayscale value. When all outputs operate at high grayscale resolution (grayscale resolution \geq 50%), the output waveform is divided into 128 segments to increase visual refresh rate. Otherwise the output waveform is divided into 64 segments at low grayscale resolution to improve output current waveform distortion. And the Δ -width correction ($\Delta \neq 0$) is used to compensate the non-ideal output current transient response.

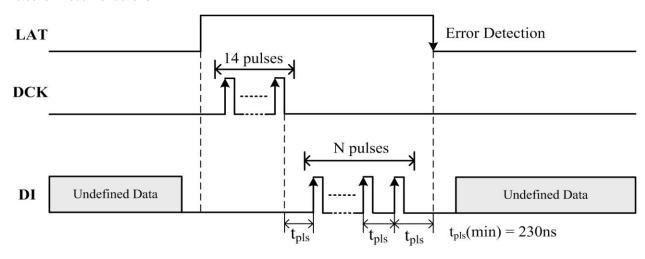
When CMD[12]="0", the output waveform operates according to APDM.

When CMD[12]="1", the output waveform is a form of non-scrambling PWM in order to drive high power LED and decrease the transient loss.



Functional Descriptions - Diagnosis and Protections

Error Detection Mode Instructions



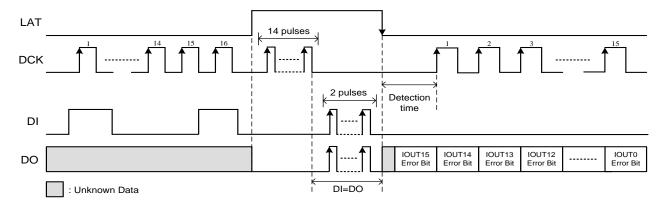
Instruction Name	Number of DI Pulses when LAT is at High	Function Description
LED Open Detection	2	Execute the LED Open Detection no matter what the image data is. The image strobe will occur at the falling edge of LAT signal.
LED Short Detection	4	Execute the LED Short Detection no matter what the image data is. The image strobe will occur at the falling edge of LAT signal.
Short-to-GND Detection	6	Execute the Short-to-GND Detection no matter what the image data is. The device will detect each IOUT pin if short-to-ground or not. The image strobe will occur at the falling edge of LAT signal.
Thermal Detection & Watchdog Result Read Out	8	The device will load thermal detection result and watchdog detection result into the 16 th and 1 st image shift register by this instruction. The thermal error occurs when the junction temperature is over 155°C.
LED Smart detection	10	Execute the LED Smart Detection no matter what the image data is. The device will detect LED Open and LED Short error simultaneously by this instruction. The image strobe will occur at the falling edge of LAT signal.
Sleep Mode	12	This instruction forces the device into Sleep Mode at the falling edge of LAT signal, then keeps GCK=high most internal circuits will shutdown. The device will be activated in the following LAT signal.
REXT Pin Detection & Watchdog Result Read Out	14	The device will load REXT detection result and watchdog detection result into the 16 th and 1 st image shift register by this instruction. The REXT error occurs when the REXT pin is short-to-GND or above 2.5V.



Error Bit

Error Bit	Symbol	Description
0		When errors, such as open errors, short errors, short-to-GND errors, and thermal errors (PreOTP), are detected in the channel, the error bit, zero, will be loaded into the corresponding image shift register.
1	Normal	In the detecting mode, the bit, one, will be loaded into the corresponding image shift register if the device is under normal operating conditions.

LED Open Error Detection



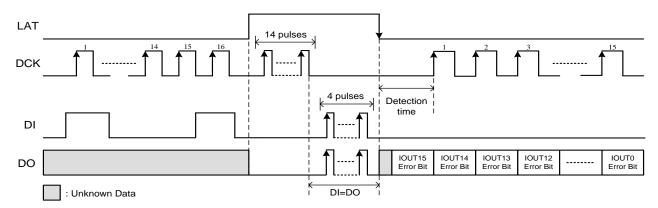
When the LED open error detection instruction is executed, it is necessary to input two DI pulses in the Error Detection Mode. No matter what the image latch data is, the output ports of AL58263 will turn on with 0.1mA for error detection. The detection time is defined from the falling edge of LAT to the rising edge of following DCK.

When the voltage of IOUT pins are below 0.3V, the error bit, zero, will be loaded into the corresponding image shift register. The error bit will be shifted out from the DO pin on the synchronization of the rising edge of DCK. And a retiming DO signal, which is similar to DI signal, appears when the LAT signal stays in high level for cascading applications.

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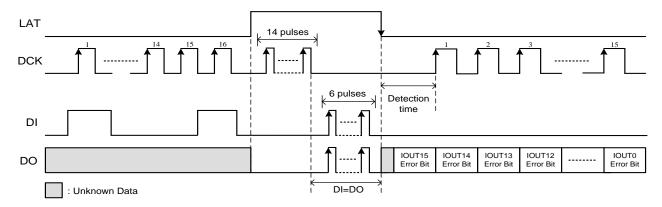


LED Short Error Detection (Output Short-to-Power Included)



When the LED short error detection instruction is executed, it is necessary to input four DI pulses in the Error Detection Mode. No matter what the image latch data is, the output ports of AL58263 will turn on with 0.1mA for error detection. The detection time is defined from the falling edge of LAT to the rising edge of following DCK. When the voltage of IOUT pins are above the threshold voltage of LED short detection determined by CMD[14:13], the error bit, zero, will be loaded into the corresponding image shift register. The error bit will be shifted out from the DO pin on the synchronization of the rising edge of DCK. And a retiming DO signal, which is similar to DI signal, appears when the LAT signal stays in high level for cascading applications.

Short-to-GND Error Detection (Output Port Leakage Included)



When the short-to-GND error detection instruction is executed, it is necessary to input six DI pulses in the Error Detection Mode. No matter what the image latch data is, this instruction forces the output ports of AL58263 to turn off for error detection. The detection time is defined from the falling edge of LAT to the rising edge of following DCK.

When the voltage of IOUT pins are below 0.3V, the error bit, zero, will be loaded into the corresponding image shift register. The error bit will be shifted out from the DO pin on the synchronization of the rising edge of DCK. And a retiming DO signal, which is similar to DI signal, appears when the LAT signal stays in high level for cascading applications.

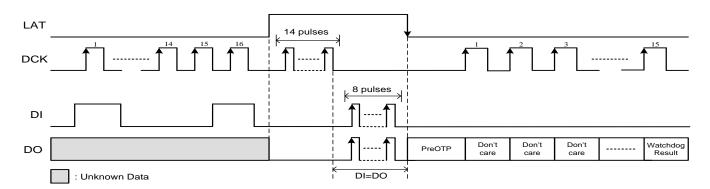
AL58263

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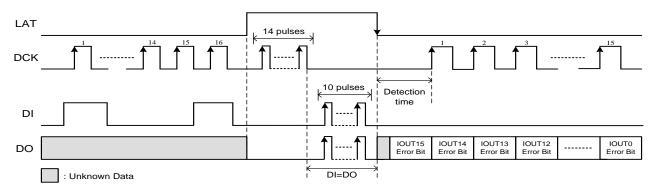
Pre-OTP Warning - Thermal Detection & Watchdog Result Read Out

The AL58263 has pre-thermal warning threshold of +155°C (typical).



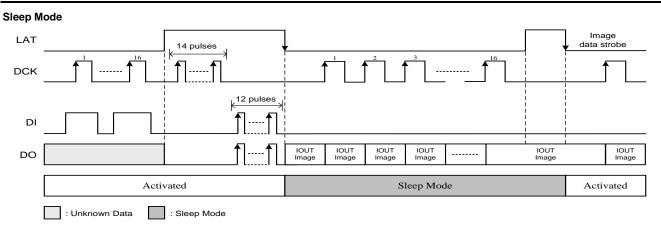
When the thermal detection instruction is executed, it is necessary to input eight DI pulses in the Error Detection Mode. AL58263 will judge of junction temperature by the instruction. If the junction temperature is over 155°C, it will lead to thermal error. The device will load the PreOTP and watchdog result into the 16th and 1st image shift register and fill out the other image shift registers with bit code, 1. The error report will be shifted out from the DO pin on the synchronization of the rising edge of DCK. And a retiming DO signal, which is similar to DI signal, appears when the LAT signal stays in high level for cascading applications. When CMD[0]=1 and GCK keeps low or high over 30ms, the watchdog result would be "0", otherwise it would be "1".

LED Smart Detection



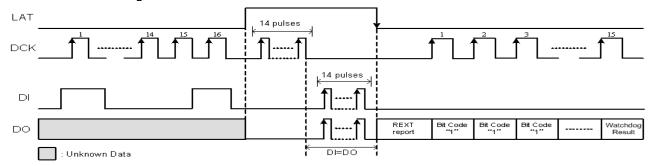
When the LED smart detection instruction is executed, it is necessary to input ten DI pulses in the Error Detection Mode. No matter what the image latch data is, the output ports of AL58263 will turn on with 0.1mA for error detection. The detection time is defined from the falling edge of LAT to the rising edge of following DCK. When the voltage of IOUT pins are below 0.3V for LED open detection or above the threshold voltage of LED short detection determined by CMD[14:13], the error bit, zero, will be loaded into the corresponding image shift register. The error bit will be shifted out from the DO pin on the synchronization of the rising edge of DCK. And a retiming DO signal, which is similar to DI signal, appears when the LAT signal stays in high level for cascading applications.





In order to power saving, AL58263 supports the sleep mode to turn off most circuits. It is necessary to input twelve pulses in the Error Detection Mode to execute this instruction and then keeps GCK=high at sleep mode. In the sleep mode, the supply current maintains < 0.1mA because most analog circuits are shutdown but the digital interface is still activated in order to receive external digital data. AL58263 will enter into sleep mode at the falling edge of this LAT signal and revive at the falling edge of following LAT signal. The serial image data will be shifted out from the DO pin on the synchronization of the rising edge of DCK. And a retiming DO signal, which is similar to DI signal, appears when the LAT signal stays in high level for cascading applications.

REXT Detection & Watchdog Result Read Out



When the REXT detection instruction is executed, it is necessary to input fourteen DI pulses in the Error Detection Mode. AL58263 will detect the voltage of REXT pin by the instruction. If the REXT pin is open or above 2.5V or below 0.15V, it will lead to REXT error. The device will load the REXT report and watchdog result into the 16th and 1st image shift register and fill out the other image shift registers with bit code, 1. The error report will be shifted out from the DO pin on the synchronization of the rising edge of DCK. And a retiming DO signal, which is similar to DI signal, appears when the LAT signal stays in high level for cascading applications. When CMD[0]=1 and GCK keeps low or high over 30ms, the watchdog result would be "0", otherwise it would be "1".



Power Dissipation

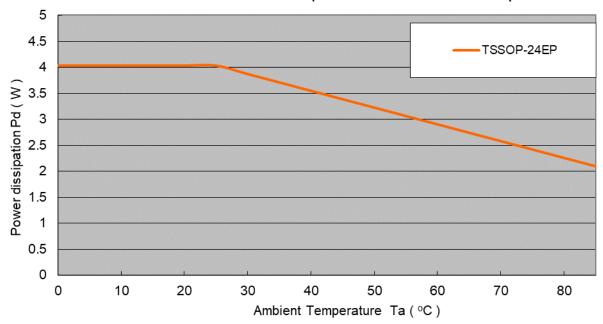
When the 16 output channels are turned on, the practical power dissipation is determined by the following equation:

$$PD(practical) = V_{DD} \times I_{DD} + Vout_{(0)} \times Iout_{(0)} \times Duty_{(0)} + \cdots + Vout_{(N)} \times Iout_{(N)} \times Duty_{(N)}$$
, where $N=1$ to 15

In secure operating conditions, the power consumption of an integrated chip should be less than the maximum permissible power dissipation which is determined by the package types and ambient temperature. The formula for maximum power dissipation is described as follows:

The PD(max) declines as the ambient temperature raises. Therefore, suitable operating conditions should be designed with caution according to the chosen package and the ambient temperature. The following figure illustrates the relation between the maximum power dissipation and the ambient temperature in the two different packages.

Maximum Power Dissipation v.s. Ambient Temperature



Design Tools (Note 16)

- AL58263 Demo Board
- Demo Board Gerber File for PCB Layout Reference

Note: 16. Diodes' design tools can be found on our website at https://www.diodes.com/design/tools/.

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Ordering Information

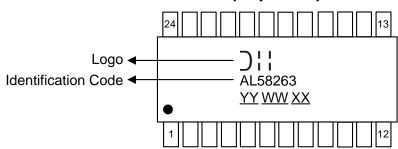


Orderable Bart Number	Package	Packing		
Orderable Part Number		Quantity	Carrier	Part Number Suffix
AL58263T24E-13	TSSOP-24EP (Type A1-B)	2,500	Tape and Reel	-13

Marking Information

(Top View - Not to Scale)

(Top View)



<u>YY</u>: Year: 24, 25, 26~ <u>WW</u>: Week: 01~52; 52

represents 52 and 53 week

XX: Internal Code

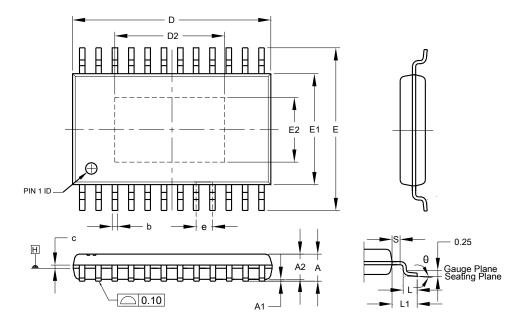
Orderable Part Number	Package	Identification Code
AL58263T24E-13	TSSOP-24EP (Type A1-B)	AL58263



Package Outline Dimensions

Please see http://www.diodes.com/package-outlines.html for the latest version.

TSSOP-24EP (Type A1-B)

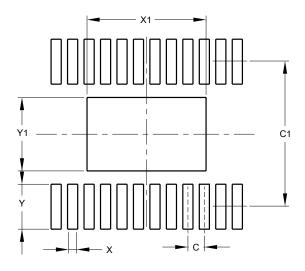


TSSOP-24EP (Type A1-B)			
Dim	Min	Max	Тур
Α		1.20	
A1	0.00	0.15	
A2	0.80	1.05	1.00
b	0.19	0.30	-
C	0.09	0.20	
D	7.70	7.90	7.80
D2	3.70	4.62	
E	6.40 BSC		<u> </u>
E1	4.30	4.50	4.40
E2	2.28	2.85	
е	0.65 BSC		
L	0.45	0.75	060
L1	1.00 REF		
S	0.20		
θ	0°	8°	
All Dimensions in mm			

Suggested Pad Layout

Please see http://www.diodes.com/package-outlines.html for the latest version.

TSSOP-24EP (Type A1-B)



Dimensions	Value (in mm)
С	0.650
C1	5.600
Х	0.400
X1	4.225
Υ	1.800
Y1	2.640

Mechanical Data

- Moisture Sensitivity: Level 3 per J-STD-020
- Terminals: Finish Matte Tin Plated Leads, Solderable per MIL-STD-202, Method 208 (3)
- Weight: 0.096 grams (Approximate)



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